Maximizing the stability radius: An LMI approach

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Abstract

Given a stabilizable linear system $E\dot{x}=Ax+Bu$ with sE-A regular, we analyze the stability robustness of the closed-loop system $(E+BK)\dot{x}=(A+BF)x+v$, obtained by proportional and derivative (PD) state feedback $u=Fx-K\dot{x}+v$. Our goal is to maximize the stability radius of the closed-loop system matrix s(E+BK)-(A+BF) over all stabilizing PD state feedback control laws. This problem turns out to be equivalent to a particular H^{∞} control problem for a generalized state-space system and reduces to a system of matrix inequalities. Under certain conditions the problem actually reduces to an LMI system. We also show how to apply these ideas to higher order dynamical systems.

1 Introduction

Subsequently the following notations will be adopted. By \mathbb{C} and $\mathbb{C}^{n\times m}$ we denote the complex field and the set of $n\times m$ complex matrices, respectively. Further, \mathbb{C}^- will stand for the open left part of the complex plane, i.e. $\{s\in\mathbb{C}: \operatorname{Re} s<0\}$, while $\mathbb{D}=\{z\in\mathbb{C}: |z|<1\}$ denotes the open unit disc.

Let us briefly recall some basic facts concerning stability radius theory. Consider a partition of the complex plane $\mathbb C$ into two disjoint sets $\mathbb C_g$ and $\mathbb C_b$, $\mathbb C = \mathbb C_g \dot{\cup} \mathbb C_b$, such that $\mathbb C_g$ is open and non-empty. Let also $E, A \in \mathbb C^{n \times n}$ such that $\Lambda(\lambda E - A) \subset \mathbb C_g$, that is, the pencil $\lambda E - A$ is $\mathbb C_g$ -stable (or, simply, stable). The two regions that are typically considered for $\mathbb C_g$ are $\mathbb C^-$ and $\mathbb D$. The unstructured complex stability radius of the pair (E,A) with respect to $\mathbb C_g$ and the perturbation $\Delta := \left[\begin{array}{cc} \Delta_E & \Delta_A \end{array}\right]$ is

$$\begin{split} r_{\mathbb{C}}(E,A,\,\mathbb{C}_g;\,\Delta) := \inf_{\Delta \in \mathbb{C}^{n \times n}} \{\|\Delta\|_2 \,:\, \exists\, \lambda \in \mathbb{C}_b \\ \text{s.t.} \ \det(\lambda(E + \Delta_E) - (A + \Delta_A)) = 0\}, \end{split} \tag{1}$$

i.e, $r_{\rm C}$ is the norm of the smallest perturbation Δ causing at least one eigenvalue of $\lambda(E+\Delta_E)-(A+\Delta_A)$

to leave the "good" region \mathbb{C}_g for \mathbb{C}_b . Here $\|\Delta\|_2 := \sigma_1(\Delta)$, where σ_1 denotes the largest singular value of Δ . Notice also that E is nonsingular since $(\lambda E - A)$ is stable. We also restrict to perturbations that do not change the infinite eigenvalues of the pencil, *i.e* $E + \Delta_E$ is nonsingular as well.

Remark 1. If $\lambda E - A$ is stable, i.e $\Lambda(\lambda E - A) \subset \mathbb{C}^-$, then

$$r_{\mathbb{C}}(E, A, \mathbb{C}^{-}; \Delta) = \left[\sup_{\omega \in \mathbb{R}} \left\| \begin{bmatrix} j\omega I \\ I \end{bmatrix} (j\omega E - A)^{-1} \right\|_{2} \right]^{-1}$$
$$= \left\| \begin{bmatrix} sI \\ I \end{bmatrix} (sE - A)^{-1} \right\|_{\infty}^{-1}. \quad (2)$$

For more details on stability radii of descriptor and higher order systems see [8], [3].

2 Problem formulation

Consider the generalized continuous-time system

$$E\dot{x} = Ax + Bu, \tag{3}$$

with $A, E \in \mathbb{C}^{n \times n}$, $B \in \mathbb{C}^{n \times m}$, such that the pencil sE-A is regular and (E,A,B) is stabilizable, i.e. $\operatorname{rank}[A-\lambda E \ B]=n \ \forall \ \lambda \in \mathbb{C} \setminus \mathbb{C}^-, \ \lambda$ finite, and $\operatorname{rank}[E \ B]=n$. Equivalently, there exist $F_0, K_0 \in \mathbb{C}^{m \times n}$ such that the pencil $s(E+BK_0)-(A+BF_0)$ has all its eigenvalues in \mathbb{C}^- . Moreover, under these conditions, let $\alpha,\beta \in \mathbb{C}$, not both zero, be such that α/β is not an eigenvalue of sE-A and $\alpha/\beta \notin \mathbb{C}^-$. Then there exist $F \in \mathbb{C}^{m \times n}$ such that $\Lambda(s(E+\beta BF)-(A+\alpha BF)) \subset \mathbb{C}^-$. More details on the generalized eigenvalue assignment problems can be found in [6]. Consider a proportional and derivative (PD) state feedback control law

$$u = Fx - K\dot{x} + v.$$

Then the system (3) becomes

$$E_K \dot{x} = A_F x + B v, \tag{4}$$

where

$$E_K := E + BK$$
, $A_F := A + BF$

and
$$\Lambda(sE_K - A_F) \subset \mathbb{C}^-$$
.

Our goal is to maximize the complex stability radius of the pair (E_K, A_F) over all PD stabilizing feedback matrices (F, K), subject to the perturbation

$$\Delta := \left[\begin{array}{ccc} \Delta_E & \Delta_A & \Delta_B \end{array} \right] \in \mathbb{C}^{n \times (2n+m)}.$$

In other words, solve

$$\sup_{F,K\in\mathbb{C}^{m\times n}} r_{\mathbb{C}}(E_K, A_F, \mathbb{C}^-; \Delta). \tag{5}$$

The complex stability radius of (E_K, A_F) with respect to Δ is defined as

$$r_{\mathbb{C}}(E_K, A_F, \mathbb{C}^-; \Delta) = \inf_{\Delta} \{ \|\Delta\|_2 : \exists \lambda \in \mathbb{C} \setminus \mathbb{C}^-$$
s. t. $\det(\lambda(E_\Delta + B_\Delta K) - (A_\Delta + B_\Delta F)) = 0 \}, (6)$

where $E_{\Delta} := E + \Delta_E$, $A_{\Delta} := A + \Delta_A$ and $B_{\Delta} := B + \Delta_B$.

Next we derive a closed formula for $r_{\mathbb{C}}(E_K, A_F, \mathbb{C}^-; \Delta)$.

Proposition 2. The complex stability radius of (E_K, A_F) with respect to $\Delta \in \mathbb{C}^{n \times (2n+m)}$ is given by

$$r_{\mathbb{C}}(E_K, A_F, \mathbb{C}^-; \Delta) = \|(C_F - sG_K)(sE_K - A_F)^{-1}\|_{\infty}^{-1}$$
 (7)

where

$$C_F := \begin{bmatrix} I \\ F \end{bmatrix}, \quad G_K := \begin{bmatrix} I \\ K \end{bmatrix}.$$
 (8)

Proof: Since $sE_K - A_F$ is stable, it follows that E_K is nonsingular. Then the eigenvalues of $sE_K - A_F$ move continuously with the perturbations Δ_E , Δ_A , Δ_B , and the eigenvalue "leaving" \mathbb{C}^- must actually lie on its boundary $\partial \mathbb{C}^-$, *i.e.* on the $j\omega$ -axis. Hence

$$r_{\mathbf{C}}(E_{K}, A_{F}, \mathbb{C}^{-}; \Delta) = r_{\mathbf{C}}(E_{K}, A_{F}, \partial \mathbb{C}^{-}; \Delta)$$

$$\begin{split} &=\inf_{\lambda\in\partial\mathbb{C}^-}\Big(\inf_{\Delta}\{\|\Delta\|_2:\det\big(\lambda(E+\Delta_E+(B+\Delta_B)K)\\ &-(A+\Delta_A)-(B+\Delta_B)F\big)=0\}\Big) \end{split}$$

$$\begin{split} &=\inf_{\lambda\in\partial\mathbb{C}^{-}}\Big(\inf_{\Delta}\{\|\Delta\|_{2}:\,\det(\lambda E_{K}-A_{F}\\ &-\left[\begin{array}{cc}\Delta_{E}&\Delta_{A}&\Delta_{B}\end{array}\right]\left[\begin{array}{c}-\lambda I\\I\\-\lambda K+F\end{array}\right])=0\}\Big) \end{split}$$

$$\stackrel{(\lambda=j\omega)}{=} \inf_{\omega \in \mathbb{R}} \left(\inf_{\Delta} \{ \|\Delta\|_2 : \det(I - \Delta \begin{bmatrix} -j\omega I \\ I \\ -j\omega K + F \end{bmatrix} \times (j\omega E_K - A_F)^{-1}) = 0 \} \right). \tag{9}$$

The last equality is due to the invertibility of $j\omega E_K - A_F$, since $\Lambda(sE_K - A_F) \in \mathbb{C}^-$.

For arbitrary $\omega \in \mathbb{R}$ one has

$$\begin{split} \inf_{\Delta}\{\|\Delta\|_2: \, \det(I-\Delta \left[\begin{array}{c} -j\omega I \\ I \\ -j\omega K + F \end{array} \right] \\ \times (j\omega E_K - A_F)^{-1}) = 0 \} \end{split}$$

$$= \| \begin{bmatrix} -j\omega I \\ I \\ -j\omega K + F \end{bmatrix} (j\omega E_K - A_F)^{-1}) \|_2^{-1}$$

$$= \| \begin{bmatrix} (-j\omega + 1)I \\ -j\omega K + F \end{bmatrix} (j\omega E_K - A_F)^{-1}) \|_2^{-1},$$

therefore

$$\begin{split} r_{\mathbf{C}}(E_K, A_F, \, \mathbb{C}^-; \, \Delta) &= \\ \inf_{\omega \in \mathbb{R}} \| \left[\begin{array}{c} (-j\omega + 1)I \\ -j\omega K + F \end{array} \right] (j\omega E_K - A_F)^{-1}) \|_2^{-1} \\ &= \left[\sup_{\omega \in \mathbb{R}} \| \left[\begin{array}{c} (-j\omega + 1)I \\ -j\omega K + F \end{array} \right] (j\omega E_K - A_F)^{-1}) \|_2 \right]^{-1} \\ &= \| \left[\begin{array}{c} (-s + 1)I \\ -sK + F \end{array} \right] (sE_K - A_F)^{-1}) \|_{\infty}^{-1}, \end{split}$$

and the proof is complete.

By substituting now $r_{\mathbb{C}}(\cdot)$ from (7) into (5), the problem can be reformulated in the following manner: Given a stabilizable system (E, A, B), solve

$$\inf_{F,K \in \mathbb{C}^{m \times n}} \left\| \begin{bmatrix} (-s+1)I \\ -sK+F \end{bmatrix} (sE_K - A_F)^{-1} \right\|_{\infty}$$

$$= \inf_{F,K \in \mathbb{C}^{m \times n}} \left\| (C_F - sG_K)(sE_K - A_F)^{-1} \right\|_{\infty}, \quad (10)$$

where C_F and G_K have been defined by (8).

3 Maximizing the stability radius: An LMI approach

Our aim is to reformulate problem (10) as a convex optimization problem, deriving F and G from the solution of some appropriate LMI's. The crucial result used in our development is the LMI version of the Bounded Real Lemma for generalized state-space systems.

Theorem 3. Consider the first order descriptor system $H(s) = D + (C - sG)(sE - A)^{-1}B$ and let $\gamma > 0$ be given. Then the following two assertions are equivalent:

- 1. sE A is stable and $||H(s)||_{\infty} < \gamma$.
- 2. There exists Y > 0 such that

$$D(Y) := - \left[\begin{array}{c} 0 \\ I \end{array} \right] \gamma^2 \left[\begin{array}{cc} 0 & I \end{array} \right] + \left[\begin{array}{c} B \\ D \end{array} \right] \left[\begin{array}{cc} B^* & D^* \end{array} \right]$$

$$\left[\begin{array}{c} A \\ C \end{array}\right] Y \left[\begin{array}{cc} E^* & G^* \end{array}\right] + \left[\begin{array}{c} E \\ G \end{array}\right] Y \left[\begin{array}{cc} A^* & C^* \end{array}\right] < 0. \tag{11}$$

3. There exists a symmetric matrix $Y \geq 0$ such that

$$R_Y := CYG^* + GYC^* + DD^* - \gamma^2 I < 0$$

and the "generalized" Riccati equation

$$AYE^* + EYA^* + BB^* - (EYC^* + AYG^* + BD^*)$$

 $\times R_V^{-1} (CYE^* + GYA^* + DB^*) = 0$

has a stabilizing solution, i.e the matrix pencil

$$sE - A + (sG - C)(EYC^* + AYG^* + BD^*) R_Y^{-1}$$
is stable.

Remark 4. The Bounded Real Lemma shows that $||H(s)||_{\infty}$ is the global minimum of the following linear objective minimization problem:

$$\inf_{\gamma, Y=Y^*} \gamma \quad \text{subject to} \quad Y > 0, \ D(Y) < 0. \tag{12}$$

According to Remark 4 and by updating formula (11) to (10) one can express $\|(C_F - sG_K)(sE_K - A_F)^{-1}\|_{\infty}$ in (10) as the global minimum of

$$\inf_{\gamma>0,\,Y=Y^{\bullet}}\,\,\gamma\quad \text{subject to}\ \, Y>0\ \, \text{and}\ \, D(Y,F,K)<0. \label{eq:constraint}$$

Here

$$D(Y, F, K) := - \begin{bmatrix} 0 & 0 \\ I_n & 0 \\ 0 & I_m \end{bmatrix} \gamma^2 \begin{bmatrix} 0 & I_n & 0 \\ 0 & 0 & I_m \end{bmatrix} +$$

$$\begin{bmatrix} I_{n} \\ 0 \\ 0 \end{bmatrix} \begin{bmatrix} I_{n} & 0 & 0 \end{bmatrix} + \begin{bmatrix} A_{F} \\ I_{n} \\ F \end{bmatrix} Y \begin{bmatrix} E_{K}^{*} & I_{n} & K^{*} \end{bmatrix} + \begin{bmatrix} E_{K} \\ I_{n} \\ K \end{bmatrix} Y \begin{bmatrix} A_{F}^{*} & I_{n} & F^{*} \end{bmatrix}. \quad (14)$$

Recall that $E_K = E + BK$ and $A_F = A + BF$. By pre and post-multiplying D(Y, F, K) by

$$U := \left[egin{array}{ccc} I & 0 & -B \ 0 & I & 0 \ 0 & 0 & I \end{array}
ight] \quad ext{and} \quad U^*,$$

then D(Y, F, K) < 0 reads as

$$U D(Y, F, K) U^* = D_2(B, \gamma) + D_1(Y, F, K) + \begin{bmatrix} 0 \\ 0 \\ I_m \end{bmatrix} (FYK^* + KYF^*) \begin{bmatrix} 0 & 0 & I_m \end{bmatrix} < 0,$$
(15)

where

$$D_2(B,\gamma) := -\begin{bmatrix} 0 & -B \\ I_n & 0 \\ 0 & I_m \end{bmatrix} \gamma^2 \begin{bmatrix} 0 & I_n & 0 \\ -B^* & 0 & I_m \end{bmatrix} + \begin{bmatrix} I_n \\ 0 \\ 0 \end{bmatrix} \begin{bmatrix} I_n & 0 & 0 \end{bmatrix}$$

and

$$D_{1}(Y, F, K) := \begin{bmatrix} A \\ I_{n} \\ F \end{bmatrix} Y \begin{bmatrix} E^{*} & I_{n} & K^{*} \end{bmatrix} + \begin{bmatrix} E \\ I_{n} \\ K \end{bmatrix} Y \begin{bmatrix} A^{*} & I_{n} & F^{*} \end{bmatrix} + \begin{bmatrix} 0 \\ I_{m} \end{bmatrix} (FYK^{*} + KYF^{*}) \begin{bmatrix} 0 & 0 & I_{m} \end{bmatrix}.$$

Assume that $M := FYF^* + KYK^* < \alpha I_m$ for given $\alpha > 0$. We relax the matrix inequality (15) to

$$D_{2}(B,\gamma) + D_{1}(Y,F,K) + \begin{bmatrix} 0 \\ 0 \\ I_{m} \end{bmatrix} (F+K)Y(F^{*}+K^{*}) \begin{bmatrix} 0 & 0 & I_{m} \end{bmatrix} < 0.$$
(16)

Since Y > 0, if (Y, F, K) is feasible for (16) then it is also feasible for (15). The above relaxation together with an appropriate change of variables allows us to rewrite the matrix inequality (16) as a LMI.

Replace F and K in (15) by introducing two new variables, $P := FY \in \mathbb{C}^{m \times n}$ and $Q := KY \in \mathbb{C}^{m \times n}$, respectively. One also has $FYK^* = PY^{-1}Q^*$, $KYF^* = QY^{-1}P^*$. With the above considerations in mind, the inequalities (15) and (16) become

$$D_{2}(B,\gamma) + D_{1}(Y,PY^{-1},Q^{-1}) + \begin{bmatrix} 0 \\ 0 \\ I_{m} \end{bmatrix} (PY^{-1}Q^{*} + QY^{-1}P^{*}) \begin{bmatrix} 0 & 0 & I_{m} \end{bmatrix} < 0.$$
(17)

and

$$D_{2}(B,\gamma) + D_{1}(Y,PY^{-1},QY^{-1}) + \begin{bmatrix} 0 \\ 0 \\ I_{m} \end{bmatrix} (P+Q)Y^{-1}(P^{*}+Q^{*}) \begin{bmatrix} 0 & 0 & I_{m} \end{bmatrix} < 0,$$
(18)

respectively. Furthermore,

$$M = FYF^* + KYK^* < \alpha I_m$$
 is equivalent to

$$\begin{bmatrix} P & Q \end{bmatrix} \begin{bmatrix} Y^{-1} & 0 \\ 0 & Y^{-1} \end{bmatrix} \begin{bmatrix} P^* \\ Q^* \end{bmatrix} - \alpha I < 0$$

$$\iff \begin{bmatrix} -Y & 0 & P \\ 0 & -Y & Q \\ \hline P^* & Q^* & -\alpha I \end{bmatrix} < 0. \tag{19}$$

Let

$$D_3(Y, P, Q, \gamma) := D_2(B, \gamma) + D_1(Y, PY^{-1}, QY^{-1}).$$

Then, since Y > 0, one also has that (18) rewritten as

$$D_{3}(Y, P, Q) + \begin{bmatrix} 0 \\ 0 \\ P + Q \end{bmatrix} Y^{-1} \begin{bmatrix} 0 & 0 & P^{*} + Q^{*} \end{bmatrix} < 0 \quad (20)$$

is equivalent to

$$\begin{bmatrix} -Y & 0 & 0 & P^* + Q^* \\ 0 & & & \\ 0 & P + Q & & D_3(Y, P, Q, \gamma) \end{bmatrix} < 0.$$
 (21)

Problem (10) can be finally reduced to the following (relaxed) optimization problem:

Given a stabilizable triple (E, A, B) and $\alpha > 0$ solve

$$\inf_{\gamma>0,Y,P,Q} \gamma \quad \text{subject to} \quad Y>0 \text{ and to} \quad (19), \quad (21).$$
(22)

If γ_0 , Y > 0, P and Q are a solution to (22), then $F := PY^{-1}, \quad K := QY^{-1}$

are a solution to the problem (10).

Note. Rather than relaxing the problem as in (16), one can also consider PD feedback control laws where $K = \beta F$, such that $\text{Re}\,\beta > 0$ (or β real and positive) and $1/\beta$ is not a generalized eigenvalue of sE - A. Let $\delta > 0$, $\delta^2 = 2\,\text{Re}\,\beta$. Replace now Q in the matrix inequality (17) by βP and obtain directly a linear matrix inequality in Y and P, which is similar to (21)

$$\begin{bmatrix} -Y & 0 & 0 & \delta P^* \\ 0 & 0 & D_3(Y, P, \beta P, \gamma) \\ \delta P & & & \end{bmatrix} < 0.$$
 (23)

These ideas can be applied to more general systems, like higher order dynamical systems. For sake of simplicity, we concentrate on systems of second order, but a similar methodology can be employed for higher order systems.

4 Robust stabilization of second order dynamical systems

Consider the following second order system

$$A_2\ddot{x} + A_1\dot{x} + A_0x = Bu, (24)$$

where $A_i \in \mathbb{C}^{n \times n}$, $i = 0, 1, 2, B \in \mathbb{C}^{n \times m}$, and the state-feedback control law

$$u = -F_2\ddot{x} - F_1\dot{x} - F_0x + v.$$

Then, the closed-loop system is given by

$$(A_2 + BF_2)\ddot{x} + (A_1 + BF_1)\dot{x} + (A_0 + BF_0)x = Bv$$

and its associated characteristic equation is

$$\det P_F(s) := \det \left(P(s) + BF(s) \right) = 0, \qquad (25)$$

where

$$P(s) = A_2s^2 + A_1s + A_0, \quad F(s) = F_2s^2 + F_1s + F_0.$$

Assuming that $P_F(s)$ is stable (i.e all the zeros of (25) lie in \mathbb{C}^-), one can now define the stability radius of P_F with respect to $\Delta = \begin{bmatrix} \Delta_{A_2} & \Delta_{A_1} & \Delta_{A_0} & \Delta_B \end{bmatrix}$, as follows:

$$r_{\mathbb{C}}(P_F, \mathbb{C}^-; \Delta) := \inf_{\Delta \in \mathbb{C}^{n \times (3n+m)}} \{ \|\Delta\|_2 : \exists \lambda \in \overline{\mathbb{C}^+}$$
s.t. $\det(P_F(\lambda) + \Delta P_F(\lambda)) = 0 \}.$ (26)

Here

$$\Delta P_F(s) = \Delta P(s) + \Delta_B F(s)$$

= $\Delta_{A_2} s^2 + \Delta_{A_1} s + \Delta_{A_0} + \Delta_B F(s)$

It can be shown that (see [3])

$$r_{\mathbb{C}}(P_F, \mathbb{C}^-; \Delta) = \| \begin{bmatrix} d(s)I_n \\ F(s) \end{bmatrix} P_F^{-1}(s)\|_{\infty}, \quad (27)$$

where $d(s) = d_2s^2 + d_1s + d_0$. Moreover, for second order systems, one gets $d_2 = d_0 = 1$, $d_1 = \sqrt{3}$.

We may now formulate the same kind of synthesis problem as that given by (10):

Given a stabilizable system (24), maximize the complex stability radius (27) over all F_0 , F_1 , F_2 . Equivalently, solve

$$\inf_{F_0, F_1, F_2 \in \mathbb{C}^{m \times n}} \| \begin{bmatrix} d(s)I_n \\ F(s) \end{bmatrix} P_F^{-1}(s) \|_{\infty}.$$
 (28)

It is easy to see that (27) can be immediately extended to linear dynamical systems of order k; accordingly, problem (28) can be also reformulated in an appropriate manner.

Some tedious manipulation yields

$$\begin{bmatrix} d(s)I_n \\ F(s) \end{bmatrix} P_F^{-1}(s) = (\widehat{C} - s\widehat{G})(s\widehat{E} - \widehat{A})^{-1}\widehat{B}, \quad (29)$$

where

$$\widehat{A} = \left[\begin{array}{cc} 0 & I_n \\ -(A_0 + BF_0) & -(A_1 + BF_1) \end{array} \right],$$

$$\widehat{E} = \left[\begin{array}{cc} I_n & 0 \\ 0 & A_2 + BF_2 \end{array} \right], \quad \widehat{B} = \left[\begin{array}{c} 0 \\ I_n \end{array} \right],$$

$$\widehat{G} = \left[\begin{array}{cc} 0 & -d_2 I_n \\ 0 & -F_2 \end{array} \right], \ \ \text{and} \ \ \widehat{C} = \left[\begin{array}{cc} d_0 I_n & d_1 I_n \\ F_0 & F_1 \end{array} \right].$$

Using the same methodology as before, by updating formula (11) to (28) and by invoking Remark 4, one obtains the equivalent problem

$$\inf_{\gamma>0,\,Z=Z^{\bullet}} \gamma \quad \text{subject to} \quad Z>0 \ \ \text{and} \ \ D(Z,F)<0. \eqno(30)$$

Clearly, D(Z, F) is D(Y) in (11) updated to (29) and Z. Let us pre and post-multiply now D(Z, F) by

$$T := \left[\begin{array}{cccc} I & 0 & 0 & 0 \\ 0 & I & 0 & B \\ 0 & 0 & I & 0 \\ 0 & 0 & 0 & I \end{array} \right] \quad \text{and} \quad T^*.$$

Then D(Z, F) < 0 is equivalent to

$$D(Z,F_0,F_1,F_2):=\left[egin{array}{c} 0\ I_n\ 0\ 0 \end{array}
ight]\left[egin{array}{cccc} 0\ I_n\ 0\ 0 \end{array}
ight]+$$

$$-\begin{bmatrix} 0 & 0 \\ 0 & B \\ I_n & 0 \\ 0 & I_m \end{bmatrix} \gamma^2 \begin{bmatrix} 0 & 0 & I_n & 0 \\ 0 & B^* & 0 & I_m \end{bmatrix} \\ + \begin{bmatrix} 0 & I_n \\ -A_0 & -A_1 \\ \frac{d_0 I_n & d_1 I_n}{F_0 & F_1} \end{bmatrix} \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{12}^* & Z_{22} \end{bmatrix} \\ \times \begin{bmatrix} I_n & 0 & 0 \\ 0 & A_2^* & -\overline{d}_2 I_n \end{bmatrix} \begin{bmatrix} 0 \\ -F_2^* \end{bmatrix}$$

$$+\begin{bmatrix} I_{n} & 0 \\ 0 & A_{2} \\ \frac{0}{0} & -d_{2}I_{n} \end{bmatrix} \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{12}^{*} & Z_{22} \end{bmatrix} \times \begin{bmatrix} 0 & -A_{0}^{*} & \overline{d}_{0}I_{n} & F_{0}^{*} \\ I_{n} & -A_{1}^{*} & \overline{d}_{1}I_{n} & F_{1}^{*} \end{bmatrix} < 0. \quad (31)$$

One can recognize now (31) as a "second order" counterpart of inequality (15). While the role of F in (15) is taken over by the block matrix $\begin{bmatrix} F_0 & F_1 \end{bmatrix}$ in (31), K is retrieved in $\begin{bmatrix} 0 & F_2 \end{bmatrix}$. If $F_2 = 0$ (which means that we restrict our analysis to control laws of the form $u = -F_1\dot{x} - F_0x + v$), then one can follow the methodology proposed in Section 3. If $F_2 \neq 0$, an additional constraint must be imposed on K, and hence on its associated variable Q. This problem is object of further research and will be addressed separately.

5 Final remarks

Several aspects are worthwhile to be emphasized. Maximizing the *complex* stability radius via state feedback is equivalent solving an appropriate state-feedback H^{∞} control problem. Such problems have been solved for instance in [1] (LMI approach) or in [4] (Riccati equation approach). A similar development can be employed to treat the discrete-time case as well.

To our knowledge, a problem which is still open is to maximize the *real* stability radius. We also mention that the related numerical aspects concerning the solution of the above mentioned (linear) matrix inequalities are under investigation.

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